

All-Optical Wavelength Conversion of Short Pulses and NRZ Signals Based on a Nonlinear Optical Loop Mirror

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Abstract—Wavelength conversion of short pulses at 10 GHz based on a nonlinear optical loop mirror (NOLM) is experimentally and numerically investigated for the case of small group velocity dispersion and walkoff between the control pulses and continuous lightwaves. Experimental and numerical simulation results show that the pulsewidths of the converted signals at different wavelengths are almost the same, and the pulsewidths are compressed when the peak power of the control pulse is smaller than a certain value. An RZ optical source containing eight wavelengths having a high sidemode suppression ratio, equal amplitudes and almost the same pulsewidths is obtained by using wavelength conversion in a NOLM consisting of a common dispersion shifted fiber. 10 Gb/s NRZ wavelength conversion based on the NOLM is demonstrated for the first time and certain conclusions in some of the references are confirmed by our experimental results.

Index Terms—Cross-phase modulation (XPM), nonlinear optical loop mirror, optical time domain multiplexing (OTDM), wavelength conversion, wavelength-division multiplexing (WDM).

I. INTRODUCTION

WAVELENGTH conversion has been suggested as a method of enhancing routing options and network properties like reconfigurability, nonblocking capability and wavelength reuse [1], [2]. To achieve all-optical wavelength conversion, various strategies have been investigated. There are mainly two kinds of media: one medium is based on semiconductor optical amplifiers (SOA's), and methods are cross-gain modulation (XGM) [3], [4], cross-phase modulation (XPM) [5], [6], and four-wave mixing (FWM) [7]–[9]; the other medium is based on fibers, and methods are FWM [10] and XPM [11]–[13] in dispersion shifted fibers (DSF's). FWM is transparent to both modulation format and signal bit-rate [7]–[10]. The use of FWM in optical networks is possible only if high conversion efficiency and high signal-to-noise ratio (SNR) can be achieved, the latter being one of the main constraints of four-wave mixing (FWM) in semiconductor optical amplifiers (SOA's) due to the high amplified spontaneous emission (ASE) [7]–[9]. Because of the SNR degradation, regeneration cannot be realized by using FWM in SOA's. However, [14], [15] have shown the regenerative capability of wavelength conversion

using an NOLM. SOA's are small and can be monolithically integrated with waveguides to form compact and polarization insensitive functional structures. But, owing to the relatively slow carrier recovery time of SOA's (~ 1 ns), nonreturn-to-zero (NRZ)-like output signals are obtained when the wavelength conversion is performed in SOA's by XGM or XPM, and this is not suitable for return-to-zero (RZ)-based networks [5]. However, nonlinear optical loop mirror (NOLM) based on silica fiber has the potential of attaining terabits per second switching operation due to the ultrafast optical nonlinearity; our experiments will demonstrate that the pulsewidths can be maintained, and even compressed when the walkoff defined in [11] between the CW waves and the control pulses is small. So, wavelength conversion based on a NOLM will be suitable for RZ based networks.

To implement practical NOLM wavelength conversion, polarization independent operation is indispensable [16]–[18]. A very simple method to overcome the polarization sensitivity of the NOLM has been experimentally demonstrated by twisting the fiber to generate circular birefringence [18]. Furthermore, experiments have demonstrated that the interference in a NOLM of a few kilometers of length is insensitive to environmentally induced changes in the linear refractive index of the fiber [16]–[19].

FWM, XPM, and XGM in SOA's have been demonstrated for NRZ wavelength conversion. But up to now, no NRZ wavelength conversion using a NOLM has been reported. In this paper, considering the effect of walkoff time between control pulses and continuous-wave (CW) light, wavelength conversion of short pulses at 10 GHz based on a NOLM will be investigated by using numerical and experimental methods.

Broadband low-noise optical sources are of vital importance for future large-capacity flexible optical network utilization of both optical time-division multiplexing (OTDM) and wavelength-division multiplexing (WDM) [20]–[22]. It is an effective method to generate short pulses by using supercontinuum, but a dispersion-flattened fiber and a special WDM filter for filtering the channels will be needed to optimize the performance [20]–[22]. We will demonstrate that an RZ WDM optical source containing eight wavelengths having the same amplitudes and nearly equal pulsewidths is successfully obtained by using wavelength conversion in a NOLM consisting of a common DSF. Furthermore, the pulsewidth of the converted pulses can be maintained and even compressed after wavelength conversion.

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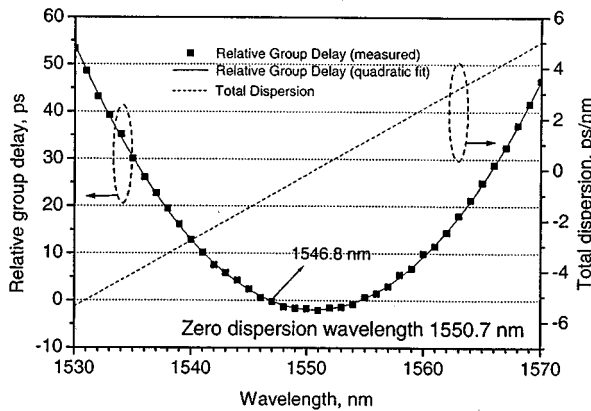


Fig. 1. Relative total group delay and total dispersion as a function of signal wavelengths for the DSF.

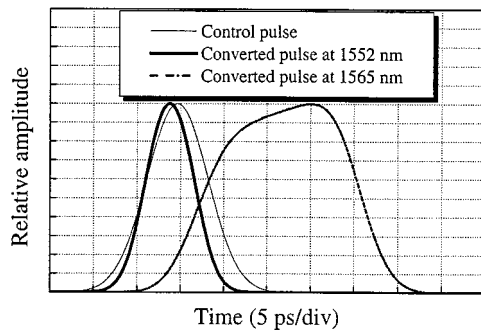


Fig. 2. Numerical simulation waveforms of converted pulses at 1552 and 1565 nm.

NRZ signals are often used because of their small bandwidth, however, up to now, no experiment of NRZ wavelength conversion based on an NOLM has been reported. We will demonstrate that also NRZ wavelength conversion based on an NOLM can be realized meaning that wavelength conversion in a NOLM is transparent to RZ and NRZ format. The conclusion in [11] that the nonlinear phase of the counterpropagating waves can be easily compensated just by adjusting the state of the polarization controller in the NOLM, and high ER of the converted signal can be obtained is confirmed by our experimental results.

This paper is divided into four sections. Section II includes some numerical simulations and experiments concerning wavelength conversion based on a NOLM for short pulses. An eight-wavelength RZ optical source having equal amplitudes and nearly the same pulsewidths is obtained by using wavelength conversion in a NOLM consisting of a common DSF. In Section III, 10-Gb/s NRZ wavelength conversion based on the NOLM is demonstrated for the first time and certain analytical results in the references are confirmed. In Section IV conclusions are given.

II. WAVELENGTH CONVERSION OF SHORT PULSES

A silica fiber based NOLM has the potential of attaining Tbit/s switching operation due to the ultrafast optical nonlin-

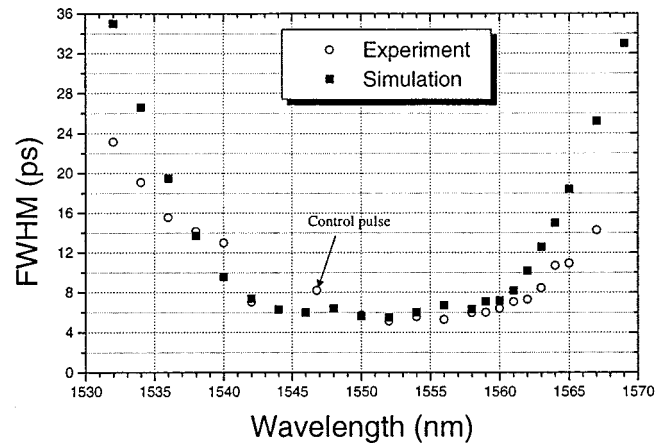


Fig. 3. FWHM pulsewidth for experiment and numerical simulation as a function of CW wavelengths. The peak power of control pulse is 150 mW, and CW power is 4 mW.

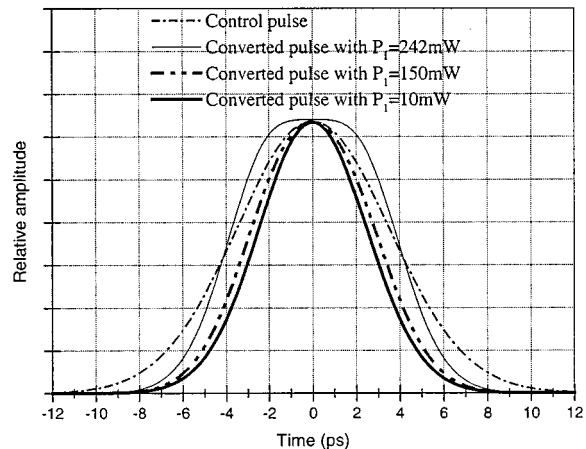


Fig. 4. Calculated pulse shapes with only XPM effect considered.

earity [23], [24], so the pulsewidth of converted pulses can be maintained. However, the walkoff between the control pulses and CW lightwave will broaden the pulsewidths of the converted signals [25]. The pulse-broadening effect can be eliminated by using dispersion flattened DSF's in NOLM's, however it is not easy to manufacture such fibers. Our experiment shows that the change in the converted pulsewidths with different converted wavelengths is very small. Furthermore, the pulsewidths of the converted signals are narrower than that of the control pulsewidth when the peak power of the control pulses, the wavelength of the control pulses and the CW lightwaves are chosen properly. The wavelength conversion based on the NOLM will be investigated using numerical and experimental methods, and the results of the numerical simulations and experiments are in good agreement. An RZ optical source containing eight wavelengths having the same amplitudes and nearly equal pulsewidths is successfully obtained by using wavelength conversion in a NOLM consisting of a common DSF. Furthermore, the pulsewidths of the converted pulses can be maintained and even compressed after wavelength conversion. In our paper, CW light waves are applied because

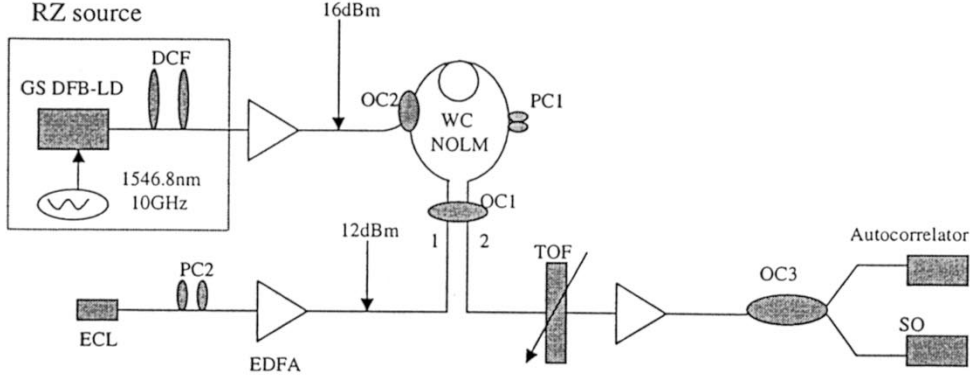


Fig. 5. Experimental setup. GS DFB-LD: gain-switched DFB laser diode. OC: optical coupler. EDFA: erbium-doped fiber amplifier. WC-NOLM: wavelength conversion-NOLM. PC: polarization controller. TOF: tunable optical filter. ECL: external-cavity laser. DCF: dispersion compensated fiber. SO: sampling oscilloscope with 32-GHz bandwidth O-E conversion.

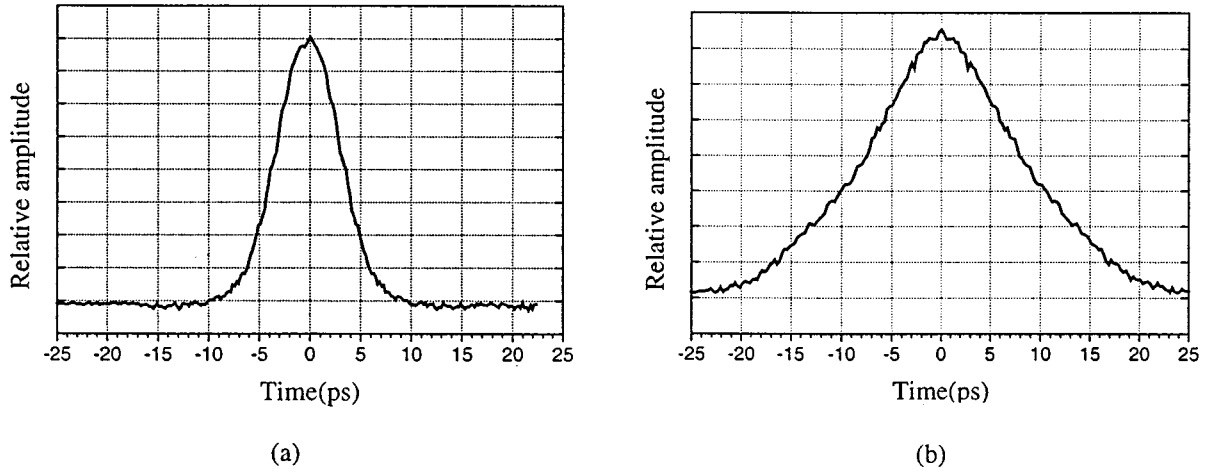


Fig. 6. Measured autocorrelator traces of converted pulses. (a) 1552 and (b) 1565 nm.

costly electronic or optical synchronization between clock pulses and control signals is then unnecessary.

A. Numerical Simulation

The propagation of the control pulses and the copropagating and counter propagating waves in the 3 km DSF is governed by the nonlinear *Schrödinger* equation (NLSE) [25], [26]. Group velocity dispersion (GVD), self-phase modulation (SPM), and XPM are incorporated into our model. When the optical pulses are converted to longer wavelengths than the wavelength of the control pulse, the Raman effect is also included in our model, and the Raman gain profile for silica fiber was approximated as a cubic spline [26]. According to [14], the XPM effect between the control pulses and the counter propagating CW lightwave can be neglected because of the small duty cycle of the control pulses and the peak power of the control pulses being far larger than their average power.

The propagation of the three waves in the fiber loop [11], control pulses, copropagating CW lightwave and counterpropagating CW lightwave can be expressed by the following coupled NLSE's [26].

Control pulses:

$$\frac{\partial A_1}{\partial z} + \frac{1}{v_1} \frac{\partial A_1}{\partial t} + \frac{i}{2} \beta(\lambda_1) \frac{\partial^2 A_1}{\partial t^2} + \frac{1}{2} \alpha A_1 = i \gamma_1 [|A_1|^2 + 2|A_2|^2] A_1 - \left[\frac{g_{12}}{2} |A_2|^2 \right] A_1. \quad (1)$$

Copropagating wave:

$$\frac{\partial A_2}{\partial z} + \frac{1}{v_2} \frac{\partial A_2}{\partial t} + \frac{i}{2} \beta(\lambda_2) \frac{\partial^2 A_2}{\partial t^2} + \frac{1}{2} \alpha A_2 = i \gamma_2 [2|A_1|^2 + |A_2|^2] A_2 + \left[\frac{g_{12}}{2} |A_1|^2 \right] A_2. \quad (2)$$

Counterpropagating wave:

$$\frac{\partial A_3}{\partial z} + \frac{1}{v_2} \frac{\partial A_3}{\partial t} + \frac{i}{2} \beta(\lambda_2) \frac{\partial^2 A_3}{\partial t^2} + \frac{1}{2} \alpha A_3 = i \gamma_2 [|A_3|^2 + 2|A_1|^2] A_3. \quad (3)$$

Here λ_i ($i = 1, 2$) is the wavelength, A_i ($i = 1, 2$ and 3) is the propagating optical field amplitude, v_i is the group velocity, $\beta(\gamma_i)$ is the linear group velocity dispersion given by $-(\lambda_i^2 D_i / 2\pi c)$ where D_i is the dispersion parameter, c is the

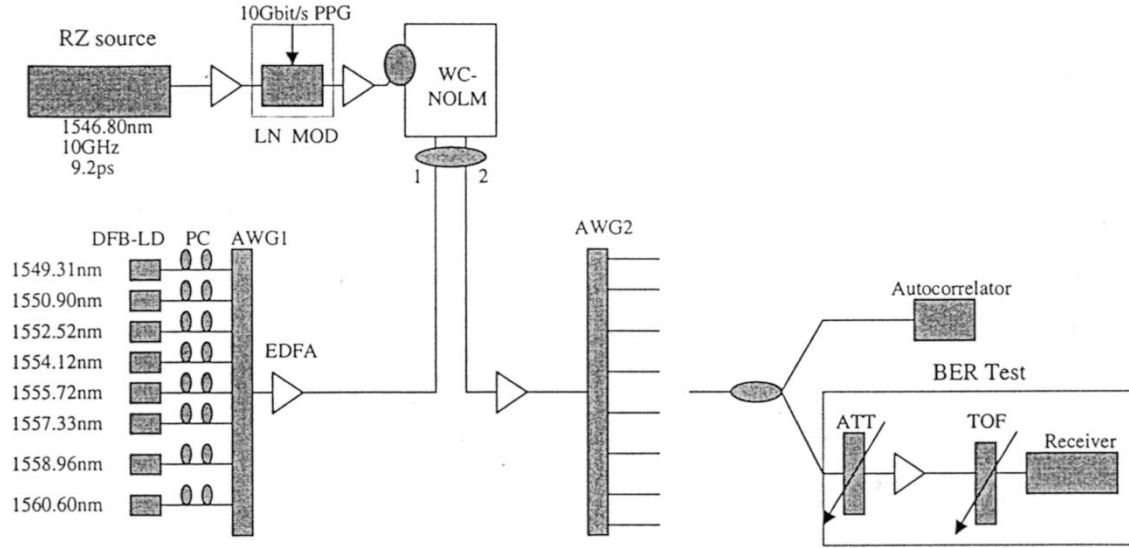


Fig. 7. Experimental setup. LN MOD: LN external modulator. AWG: arrayed waveguide grating. PPG: pulse pattern generation. ATT: attenuator. BER: bit-error rate detector. WC-NOLM: wavelength conversion NOLM. Note: The LN modulator is not used for the experiment RZ pulse conversion.

speed of light, γ_i is the nonlinear coefficient defined by $\gamma_i = n_2 \omega_i / c A_{\text{eff}} = 2\pi c n_2 / c A_{\text{eff}} \lambda_i$, where the nonlinear index coefficient $n_2 = 2.67 \times 10^{-20} \text{ m}^2/\text{W}$, and A_{eff} is the effective fiber core area. $A_{\text{eff}} = 50 \mu\text{m}^2$. g_{12} is the Raman gain coefficient between the control pulses and the copropagating CW lightwave. α is the loss of the DSF, and we assume that it is independent of wavelength. $\alpha = 0.25 \text{ dB/km}$. The FWHM pulsewidth, center wavelength, repetition frequency and peak power of control pulses are 8.2 ps, 1546.8 nm, 10 GHz, and 150 mW, respectively. The shape of the control pulse is Gaussian. The wavelength of the CW lightwave is changed from 1533 nm to 1569 nm while the average powers of copropagating and counter propagating CW lightwaves are both 4 mW. The DSF total dispersion and relative group delay are shown in Fig. 1.

Fig. 2 shows numerical simulation waveforms of the converted pulses at the wavelengths of 1552 and 1565 nm. The pulse shape becomes asymmetric and broad because of a large walkoff when the wavelength of the CW lightwave is 1565 nm [26]. Fig. 3 shows the FWHM pulsewidths (experimental and calculated) of the converted signals as a function of the CW wavelengths. From Fig. 3 we can see that the experimental pulsewidths of the converted pulses with wavelengths from 1542 to 1563 nm are compressed compared to control pulse. Within this wavelength range, there is a small walkoff and small dispersion, so XPM is the main effect. Assuming that the shape of the control pulse is Gaussian, when only XPM effect is considered, the power of the converted pulses can be expressed by the following equation [14], [15]:

$$P_{\text{out}} = \frac{P_2}{2} \left[1 - \cos \left(2\gamma_1 P_1 L \exp \left(-\frac{t^2}{t_c^2} \right) \right) \right]. \quad (4)$$

Here L is the length of DSF, P_1 and P_2 are the peak power of control pulses and the power of the CW lightwave, respectively. t_c is the pulsewidth of the control pulses which is related to the FWHM pulsewidth by $\text{FWHM} \approx 1.665 t_c$.

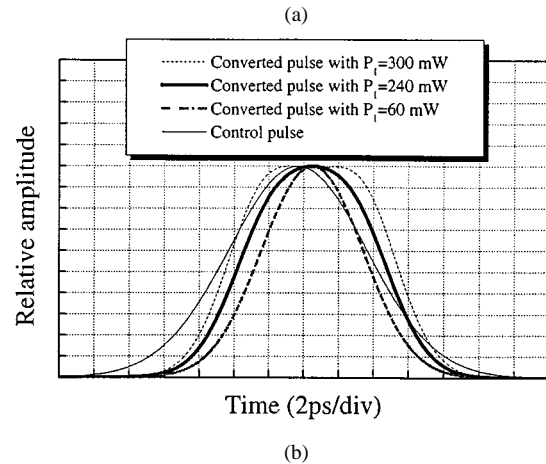
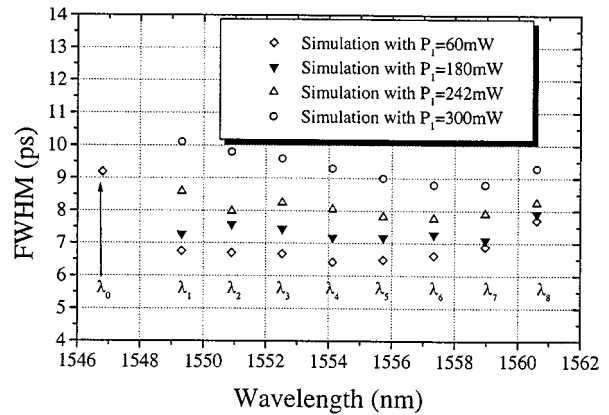


Fig. 8. Numerical simulation results with control pulsewidth of 9.2 ps and different control power (P_1). (a) FWHM at eight wavelengths and (b) the λ_1 waveform for different control peak powers.

With only XPM effect considered, Fig. 4 shows the waveforms obtained from (4). It is clearly seen that the converted pulsewidth depends on the peak power of the control pulses.

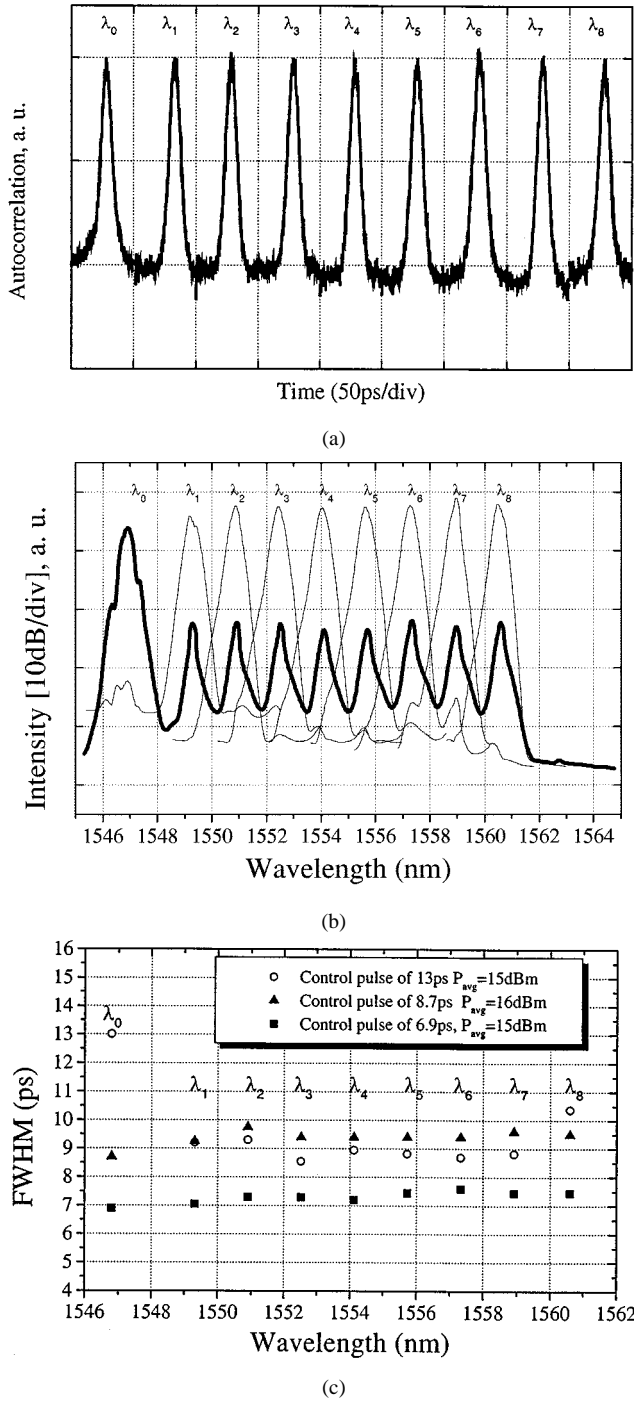


Fig. 9. Experimental results, λ_0 represents control pulse. (a) Autocorrelation waveform for control pulse of 8.7 ps. (b) Optical spectrum for control pulse of 8.7 ps, the thick and thin curves represent the optical spectrum from Port 2 before AWG2 filter and after AWG2 filter, respectively. (c) Pulse FWHM for different control pulse widths and the average powers (P_{avg}).

When the peak power of the control pulses is smaller than 242 mW, the converted pulses are compressed. When the peak power of the control pulse is larger than 242 mW, the converted pulses will be split and the pulse widths will be broadened. With only XPM effect considered, the peak power of the control pulse for the conversion efficiency [19] of 1 is 242 mW. It should be pointed out that the peak power of the control pulse for conversion efficiency of 1 will be larger than 242 mW if also

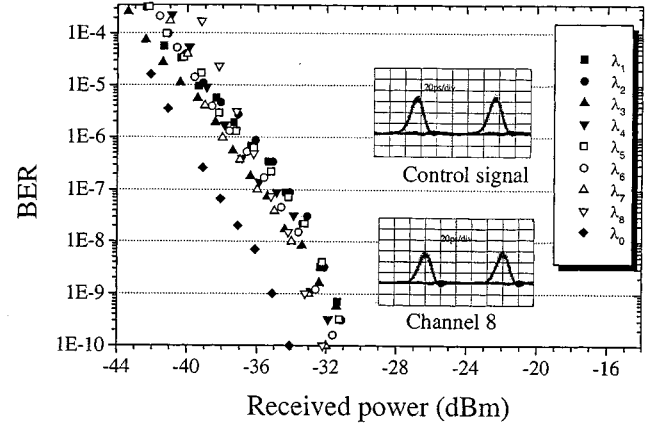


Fig. 10. BER versus received power with control pulse width of 8.7 ps, λ_0 represents control signal.

other effects are taken into account [19]. Because XPM effect is independent of the converted wavelengths, the pulse widths at different converted wavelengths will be the same when only XPM effect is considered. Our numerical simulation shows that SPM and Raman effects only have a small influence on the pulse widths of the converted pulses. When the wavelength of the CW lightwave is increased to a certain value, dispersion effect and walkoff effect can no longer be ignored. In such cases the converted pulse widths will be quickly broadened as shown in Fig. 3.

B. Wavelength Conversion Experiments with Single Wavelength Pulses

The experimental diagram is shown in Fig. 5. The control laser is a 10-GHz, 1546.8-nm (λ_0) gain-switched distributed feedback laser diode (DFB LD) that generates 8.2-ps (FWHM) pulses after compression in a dispersion compensated fiber (DCF). The on-off ratio [21] between maximum and minimum transmission of the NOLM is 30 dB without the control signal when the state of polarization controller 1 is varied. The CW lightwave is generated by a tunable external cavity laser. The control signal is coupled into the NOLM using a wavelength independent 3 dB optical coupler (OC2). Polarization controller 1 in the loop is used to obtain best performance of the converted pulses. A 1.6-nm bandpass tunable optical filter (TOF) at the output of the NOLM is used to suppress the control pulses. The average power of the control pulses into the 3 dB OC2 is 16 dBm, and the power of the CW lightwave into OC1 is 12 dBm. All OC's are 2×2 couplers with power coupling ratios of 50:50%. The length of the DSF in the NOLM for wavelength conversion (WC-NOLM) is 3 km. We measured the total group delay and dispersion of the DSF in the WC-NOLM. The results are shown in Fig. 1. It is a common DSF with zero dispersion wavelength of 1550.7 nm and dispersion slope of $0.08 \text{ ps/nm}^2/\text{km}$.

The pulse traces are measured by an autocorrelator and the measured results are shown in Fig. 6. Assuming that the waveforms of the converted pulses have Gaussian shapes, then the FWHM pulse widths are equal to the second harmonic generation (SHG) trace pulse widths divided by 1.414. Of course, the

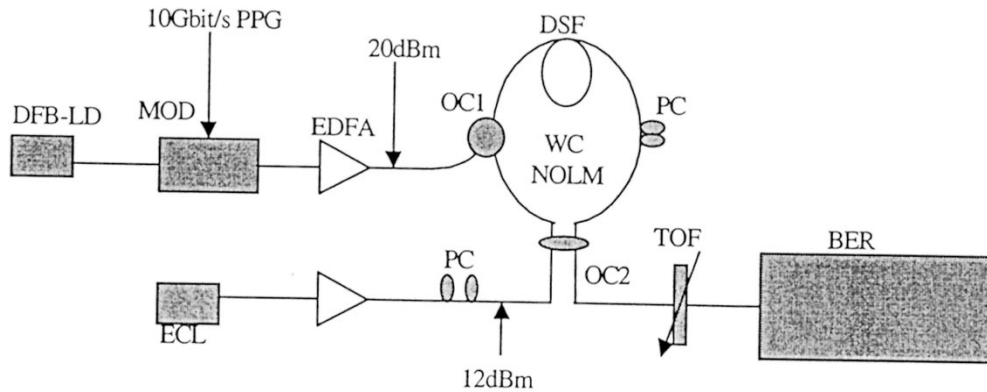


Fig. 11. Experimental Setup.

results will not be accurate when the pulse shape of the converted pulse is not Gaussian. So there is a considerable difference between the numerical simulation and experimental results when the wavelength separation between the control pulse and the CW lightwave is large. However, the numerical simulation and experimental results are in good agreement when the wavelength separation is small. Some typical autocorrelator traces are shown in Fig. 6. Because of the large walkoff, the pulses at 1565 nm are broadened severely.

C. Experiments with Wavelength Conversion into Multiwavelength Pulses

Broadband low-noise optical sources are of vital importance for future large-capacity flexible optical network utilization of both OTDM and WDM [20]–[22]. It is an effective method to generate short pulses by using supercontinuum, but a dispersion-flattened fiber and a special WDM filter for filtering the channels will be needed [20]–[22].

From Fig. 3, we can see that almost equal pulsewidths at different wavelengths are obtained when the wavelength separation between the control pulses and the CW lightwave is small. In this section, simultaneous RZ pulses at eight different wavelengths are realized by using wavelength conversion in an NOLM consisting of a common DSF. The NOLM for wavelength conversion is the same as that in the above section. The converted wavelengths of the eight signal pulses are in agreement with the ITU-T proposal (200 GHz spacing) and a commercial arrayed-waveguide grating (AWG) is used to realize demultiplexing in the frequency-domain. The converted pulses at the different wavelengths have nearly equal widths and same amplitudes.

The experimental setup is shown in Fig. 7. The outputs of eight commercial DFB lasers are multiplexed in a commercial AWG having a wavelength spacing of 1.6 nm (it can be substituted by an 8×1 optical coupler). These lasers operate according to the ITU-standardized wavelength proposal. The operating wavelengths are 1549.31 nm (λ_1), 1550.90 nm (λ_2), 1552.52 nm (λ_3), 1554.12 nm (λ_4), 1555.72 nm (λ_5), 1557.33 nm (λ_6), 1558.96 nm (λ_7) and 1560.60 nm (λ_8). The combined signals of the eight channels are amplified to an average power of 18 dBm, and injected into the NOLM from Port 1. The control laser is a 10 GHz, 1546.8 nm (λ_0) gain-switched DFB LD

that generates 6.0–13.0 ps (FWHM) pulses after compression in a DCF. In order to get different pulsewidths of the converted signal, the average power of the control pulses is adjustable (13 to 18 dBm). The RZ pulses at eight wavelengths are realized simultaneously by using wavelength conversion in the NOLM. Fig. 8 shows the numerical simulation results with the control pulsewidth of 9.2 ps and different peak power of the control pulses. From Fig. 8(a) we can see that the converted pulsewidth are nearly independent of wavelengths for each control power. Fig. 8(b) shows that the converted pulsewidths are dependent on the control power.

Fig. 9 shows the experimental results. Equal amplitudes, nearly equal pulsewidths and larger than 30-dB sidemode suppression ratios at different wavelengths are obtained. From Fig. 9(c) we can see that the pulsewidths of the converted pulses can be made narrower or a little wider than that of the control pulses by changing the power and the pulsewidths of the control pulses.

After adding the LN modulator, and modulating the optical pulses with a pseudorandom bit sequence (PRBS) of $2^{31} - 1$, BER's of the eight 10 Gb/s channels are measured. The average power of the control pulses is 15 dBm. The BER curves in Fig. 10 show no error floors and the measured power penalties for $\text{BER} < 10^{-9}$ are all within 4 dB. The power penalties are mainly attributed to reduced signal-to-noise ratio (SNR). From Fig. 9(b) we can see that the SNR of the converted signals is over 10 dB smaller than that of the control signals. Increasing the input power of the CW light into the NOLM is beneficial for SNR improvement of the converted signal, and for reducing the power penalty. Our experiment has demonstrated that the power penalty is very small and even negative when only one channel of CW light with an average power of 14 dBm is operated [12]. The un-flattened gain of the EDFA leads to different penalties for the different channels.

III. NRZ WAVELENGTH CONVERSION

There are two main difficulties with NRZ wavelength conversion. One is a high peak power requirement for the control pulse to obtain a high conversion efficiency. However, the duty cycle of NRZ signals is 1, and the peak power of the NRZ signals is only twice the average power. So it is difficult to get an NRZ

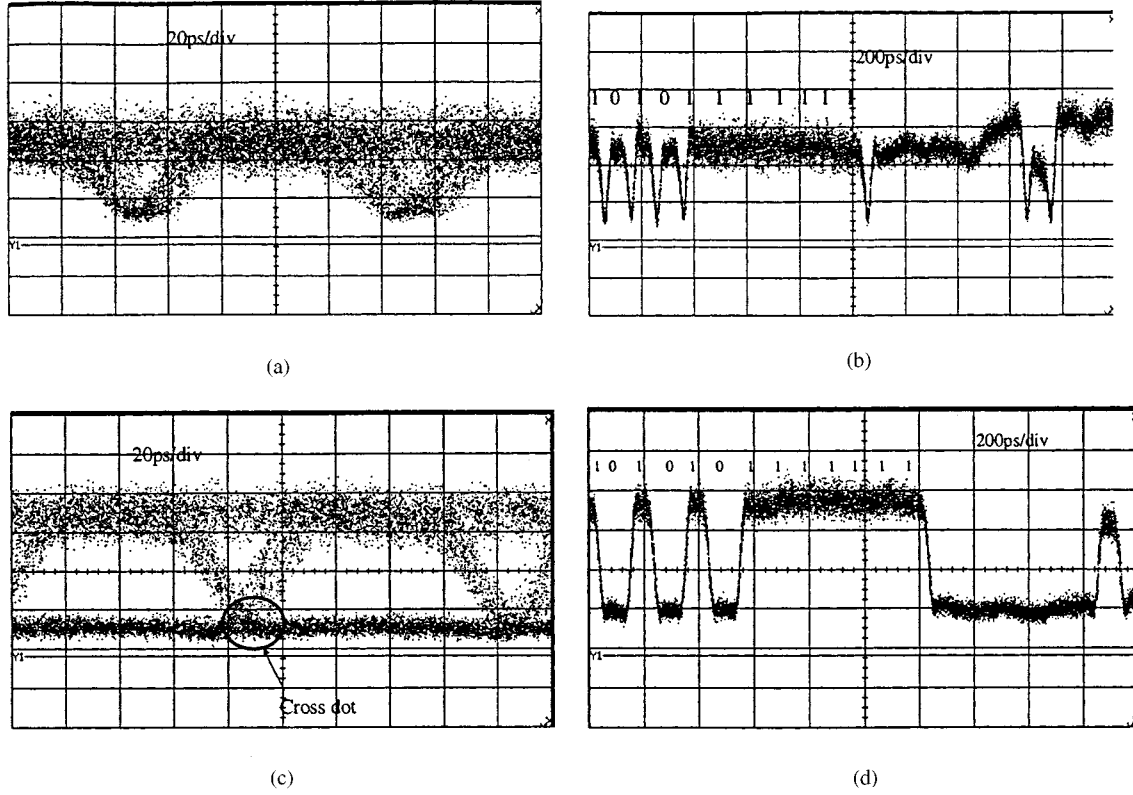


Fig. 12. (a) and (b) Eye diagram and waveform of the converted signal at 1546 nm without adjusting the polarization controller in the NOLM, respectively. (c) and (d) Eye diagram and waveform the converted signal at 1546 nm with adjusting the polarization controller in the NOLM.

pulse with a high peak power because of the finite average power of EDFA output. Another problem is that the counter propagating signal waves will generate a large unwanted nonlinear phase shift, which will lead to a poor extinction ratio (ER) of the converted pulses. Reference [11] has shown that the unwanted nonlinear phase shift of the counter propagating wave can be compensated by adjusting the state of polarization controller in the NOLM. Our experiment will confirm this observation, and 10-Gb/s NRZ wavelength conversion based on a NOLM is achieved for the first time.

A. Experiment

The experimental setup is shown in Fig. 11. The WC-NOLM is the same as that in Fig. 7. The control signal with the center wavelength of 1556 nm is externally modulated by a LN modulator to 10 Gb/s using a PRBS of $2^{11} - 1$. The control signal is coupled into the NOLM using OC1. The CW lightwave from an external cavity laser is injected into the NOLM using OC2. A 1.6-nm bandpass filter at the output of the NOLM is used to suppress the control signal. The average power of the control signal into OC1 is 20 dBm, and that of the CW lightwave into OC2 is 12 dBm. In this experiment, the range of the tunable optical filter (TOF) is from 1530 to 1561 nm.

B. Results and Discussions

In order to obtain maximum nonlinear phase shift between the control signal and the co-propagating wave, we assume that

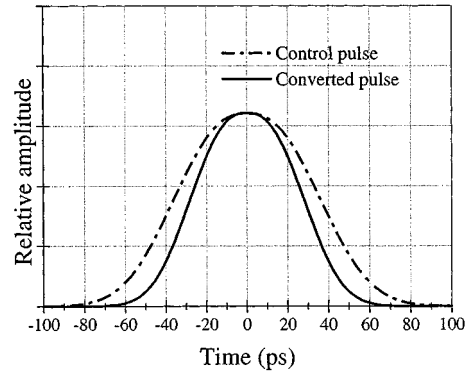


Fig. 13. Calculated pulse waveform with only XPM considered and $P_1 = 100$ mW.

they have parallel polarization. According to Ref. [11], ER of the converted signal can be written as

$$ER = \frac{\sin^2((\hat{\Phi}_{co} - (\Phi_{cn} - \Phi_{cn,0}))/2)}{\sin^2((\Phi_{cn} - \Phi_{cn,0})/2)}. \quad (5)$$

Here $\hat{\Phi}_{co}$ and Φ_{cn} denote the maximum phase shift of the copropagating wave and phase shift of the counterpropagating wave produced by XPM from the control signal, respectively. $\Phi_{cn,0}$ denotes the fixed phase shift of the counter propagating signal caused by the polarization controller in the NOLM.

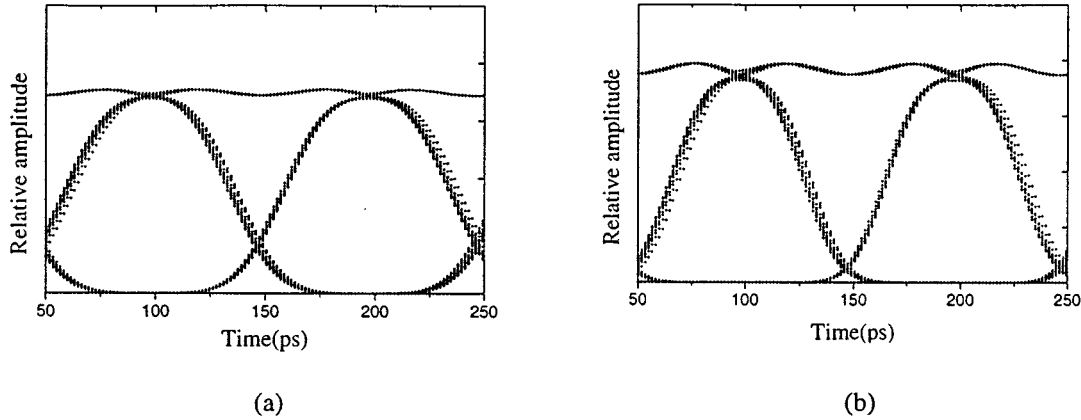


Fig. 14. Eye diagrams of numerical simulation. (a) Control signal (1556 nm). (b) Converted signals (1546 nm).

Without considering GVD and SPM in the DSF, and assuming no walk-off between control pulses and signal pulses, and if polarization between counter propagating CW lightwave and control signal is parallel, then $\Phi_{cn,0} = 0$. $\hat{\Phi}_{co}$ can be expressed as $2\gamma_1 P_1 L$ [11], [14], and Φ_{cn} can be expressed as $\gamma_1 P_1 L$ when there is equal distribution of “1”s and “0”s in the control pulse train [11], [14], here P_1 is the peak power of the control pulse, L is the length of the DSF and γ_1 is the nonlinear coefficient of the fiber. In this case, according to (5), the ER of the converted signal is one, and Fig. 12(a) and (b) shows, respectively, the optical eye diagram and the waveform of the converted signal for the center wavelength of the control signal of 1556 nm. We can see that there is nearly no difference between the “1”s and “0”s, and the eye is almost closed.

According to (5), in order to obtain high ER of the converted signal, we must find a proper $\Phi_{cn,0}$ to make $\Phi_{cn} = \Phi_{cn,0}$. It means that $\Phi_{cn} = \beta\gamma_s P_0 L/2 = \Phi_{cn,0}$. β represents the XPM parameter between the counter propagating wave and the control signal and it ranges from 2/3 to 2 [26]. Fig. 12(c) and (d) shows, respectively, the eye diagram and the waveform of the converted signal after adjusting the polarization state of polarization controller in the NOLM. We can see that there is a high ER of the converted signal. It demonstrates that the nonlinear phase shift of the counterpropagating wave can be compensated by adjusting the state of polarization controller in the NOLM.

Observing Fig. 12(c), we can see that the cross dot is moved to a very deep position. The reason is that the width of “mark” is compressed. Assuming the control pulse to be a super Gaussian pulse with a sharpness figure of 1.436 [26], the nonlinear phase shift of counterpropagating wave is compensated by adjusting polarization controller in the NOLM, and only considering XPM effect, the output of the converted pulse can be expressed as

$$P_{out} = P_2 \left[1 - \cos \left(2\gamma_1 P_1 L \exp \left(- \left(\frac{t}{t_c} \right)^{2.872} \right) \right) \right]. \quad (6)$$

The meaning of symbols in (6) is the same as that in (4). $t_c = 44.6$ ps. The waveforms of the pulses obtained from (6) are shown in Fig. 13. It is clearly seen that the converted pulse is compressed. Fig. 14 shows the numerical simulation results

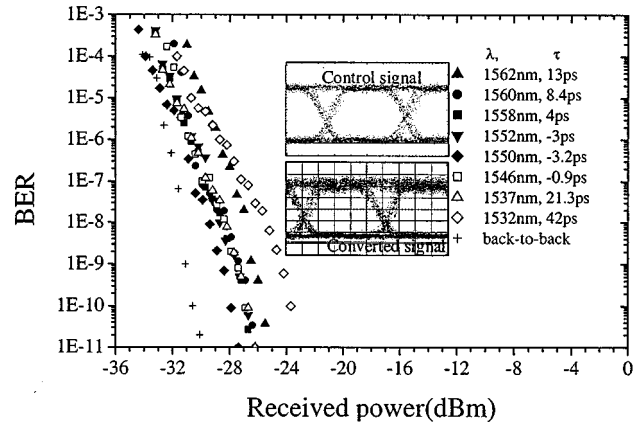


Fig. 15. BER versus received power. τ represents the walkoff time between the control signal and CW lightwave. The insets are the eye diagrams of the control signal and the converted signal (1560 nm).

with XPM, SPM, GVD and Raman effects considered. The parameters for the numerical simulation are the same as those in the above experiment. From Fig. 14, we can see that the cross dot is moved to a low position and that is in good agreement with Fig. 12(c).

Fig. 15 shows the BER performance of the control signal (back-to-back) and the converted signals at different center wavelengths. The insets are the eye diagrams of the control signal and the converted signal (1560 nm). The maximum and minimum power penalty of the converted signals is 7 dB and 2.6 dB, respectively. We can see that there is no error floor for the converted signals at the different wavelengths. Fig. 16 shows the eye diagrams of numerical simulation of converted signals at 1532 nm and 1562 nm without considering ASE noise of EDFA's. Our numerical simulation shows that when the wavelengths of the CW light is chosen to be from 1537 to 1560 nm, the converted eye diagrams have no obvious difference, because the walkoff time and dispersion within this wavelength range are not very large; XPM is the main effect, and XPM is independent of wavelength. So, the power penalties of the converted signals should be equal. In our experiment, the polarization controller in the NOLM plays an important role because

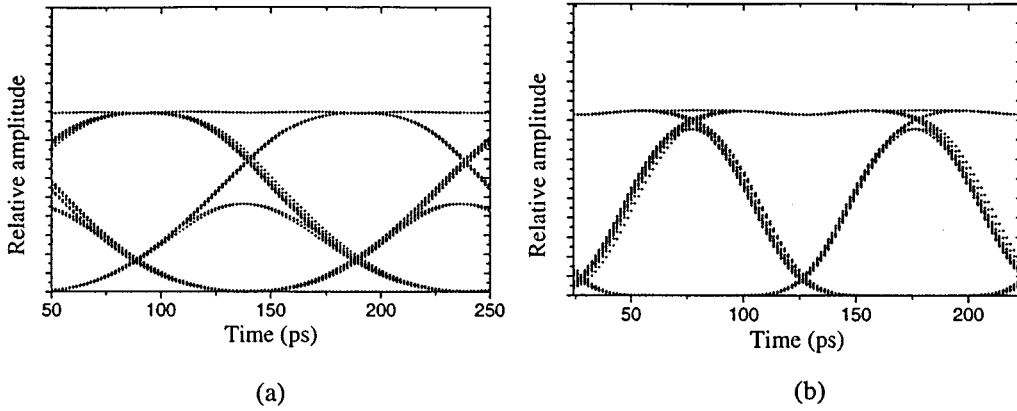


Fig. 16. Eye diagrams of numerical simulation of converted signals without considering ASE noise of EDFA's (a) 1532 and (b) 1562 nm.

ER improvement is realized by adjusting the polarization state of the polarization in the NOLM. However, considering the precision in adjusting the polarization controller in the NOLM and the fluctuation in the states of polarization in the fibers, an error in the power penalty might exist at each BER curve. Therefore, a maximum difference of approximately 1 dB in power penalty between different channels is shown in Fig. 15 when the wavelengths are chosen to be from 1537 nm to 1560 nm.

When the wavelength of the CW light is chosen to be 1532 nm, we can see from Fig. 15 that the converted signals have about 7 dB power penalty. We believe the power penalty is mainly caused by the following two factors. First, large walkoff will lead to an obvious intersymbol interference (ISI), which can be observed in Fig. 16(a), and the ISI will in turn lead to some power penalty. Second, because all EDFA's in our experiment are common EDFA's, we can see from Fig. 17 that the noise figure of the EDFA used to amplify the CW light is very large when the wavelength is 1532 nm and which large ASE noise of EDFA leads to the converted signals having a small SNR, in turn will increase the power penalty.

When the CW light wavelength is 1562 nm, from Fig. 16(b), no obvious ISI can be observed, but 1562 nm is a little beyond the range of the TOF. So the converted signals after the TOF are distorted and this leads to a little larger penalty.

The optical receiver output depends on the input pulse shape and pulsewidth, so pulse shapes and pulsewidths of the converted signals will also have some effect on the power penalty. When properly increasing the length of the DSF or the output power of EDFA3, the power penalties of the converted signals will be reduced. In order to obtain the conversion efficiency of 1, the requirement for the peak power of the control signals should be 242 mW (only XPM effect considered). However, in our experiment, we can only obtain 100 mW for the peak power of the control signals.

IV. CONCLUSION

Wavelength conversion of short pulses at 10 GHz based on an NOLM has been experimentally and numerically investigated. With small walkoff between the control pulses and the CW light-

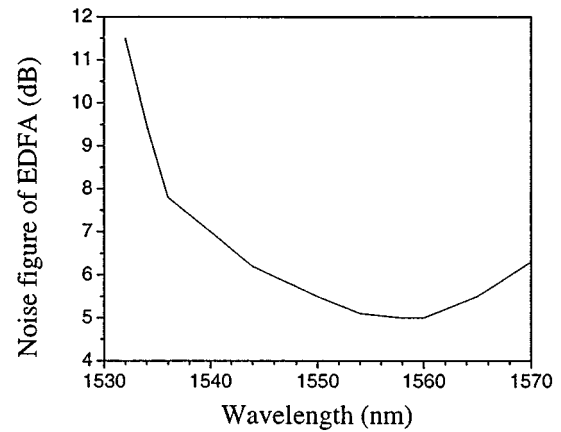


Fig. 17. Measured noise figure of EDFA used to amplify the CW light with an input power of -3 dBm.

waves and small GVD, the pulsewidths of the converted pulses are dependent on the power of the control pulses. Furthermore, the converted pulsewidths at different wavelengths are nearly the same. Because the pulsewidths of the converted signals can be maintained and even compressed, the converted signals will be suitable for RZ based networks.

An eight-wavelength RZ optical source having high side-mode suppression ratios, equal amplitudes and almost the same pulsewidths using wavelength conversion in a NOLM consisting of a common DSF is demonstrated. The converted wavelengths of the eight channels adhere to the ITU-T proposal, and the demultiplexing in the frequency-domain is realized by a common commercial AWG.

NRZ wavelength conversion at 10 Gb/s based on a NOLM has been demonstrated. We have experimentally confirmed that the nonlinear phase shift of the counter propagating wave can be compensated by adjusting the state of polarization controller in the NOLM and converted signals with high ER can be obtained.

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